

# AC CONDUCTANCE AND NON-SYMMETRIZED NOISE AT FINITE FREQUENCY IN QUANTUM WIRE AND CARBON NANOTUBE

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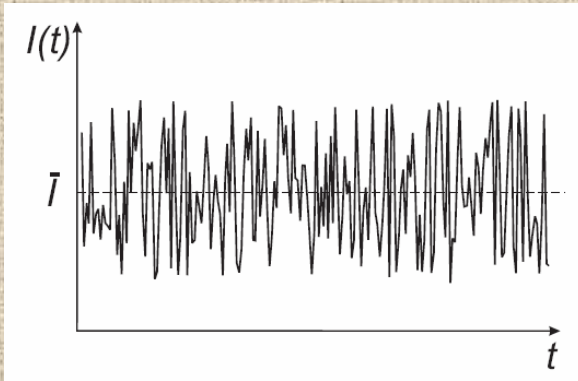
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# CURRENT FLUCTUATIONS



## ORIGINS OF NOISE

- High temperature : Johnson-Nyquist noise
- High voltage : Shot noise
- High frequency : Quantum noise
- We neglect the 1/f noise

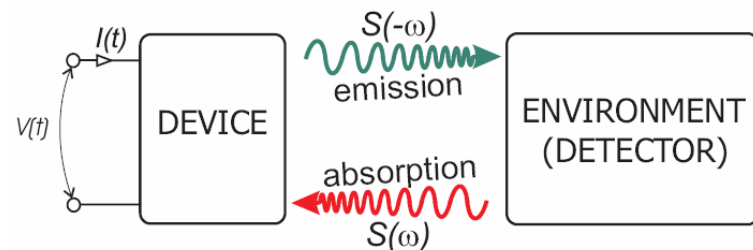
## ZERO FREQUENCY AND ZERO TEMPERATURE: SHOT NOISE

⇒ Schottky relation  $S^+(\omega=0) = e * |\langle \delta j \rangle|$  where  $S^+(\omega) = FT \left\{ \frac{1}{2} [\langle \delta j(0) \delta j(t) \rangle + \langle \delta j(t) \delta j(0) \rangle] \right\}$   
symmetrized noise

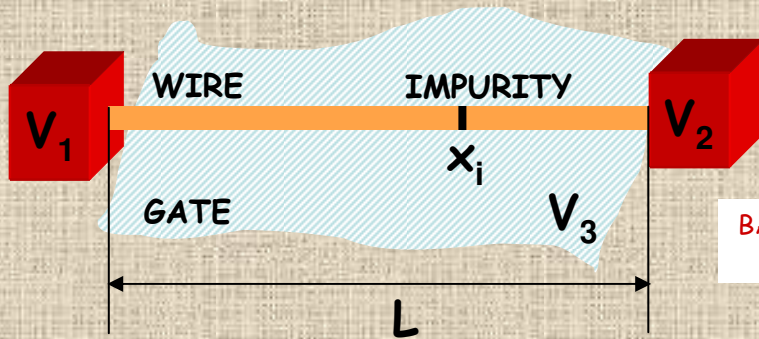
## HIGH FREQUENCY NOISE MEASUREMENTS

- High-frequency measurement in a diffusive wire    1 – 20GHz    *SCHOELKOPF et al., PRL (1997)*
- On-chip detection using SIS junction                      → 100 GHz    *DEBLOCK et al., Science (2003)*
- Direct measurement in a QPC                                      4 – 8GHz    *ZAKKA-BAJJANI et al., PRL (2007)*

What is measured is  $S(\omega) = \int dt e^{i\omega t} \langle \delta j(0) \delta j(t) \rangle$   
non-symmetrized noise



# THE SYSTEM



## MODEL

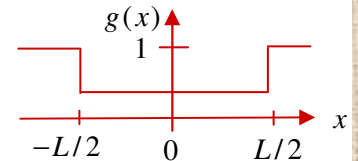
$$H = H_0 + H_B + H_V$$

$$H_0 = \frac{\hbar v_F}{2} \int_{-\infty}^{\infty} dx \left[ \Pi^2 + \frac{1}{g^2(x)} (\partial_x \Phi)^2 \right]$$

$$H_B = \lambda \cos \left[ \sqrt{4\pi} \Phi(x_i, t) + 2k_F x_i \right]$$

$$H_V = - \int_{-\infty}^{\infty} \frac{dx}{\sqrt{\pi}} eV(x) \partial_x \Phi(x, t)$$

COULOMB INTERACTIONS PARAMETER



BACKSCATTERING AMPLITUDE

## WE CALCULATE PERTURBATIVELY

□ The non-symmetrized noise:  $S_{nm}(\omega) = \int dt e^{i\omega t} \langle \delta j_m(0) \delta j_n(t) \rangle$  where  $n, m = 1, 2, 3$

□ The AC conductance:  $G_{nm}(t-t') = \frac{\delta I_n(t)}{\delta V_m(t')} \Big|_{V_m=0}$  where  $\begin{cases} \delta I_n(t) = \langle \delta j_n(t) \rangle \\ V_n(t) = v_n \cos(\omega t) \end{cases}$

## RESULT

$$S_{nm}(\omega) = S_{nm}^+(\omega) - \hbar \omega \text{Re}[G_{nm}(\omega)]$$

with  $S_{nm}^+(\omega) = \frac{1}{2} \int dt e^{i\omega t} \langle \delta j_m(0) \delta j_n(t) + \delta j_m(t) \delta j_n(0) \rangle$  **symmetrized noise**

## GENERALIZED KUBO-TYPE FORMULA

$$S_{nm}^-(\omega) = S_{nm}(\omega) - S_{nm}(-\omega) = -2\hbar \omega \text{Re}[G_{nm}(\omega)]$$

**SAFI AND SUKHORUKOV (unpublished)**

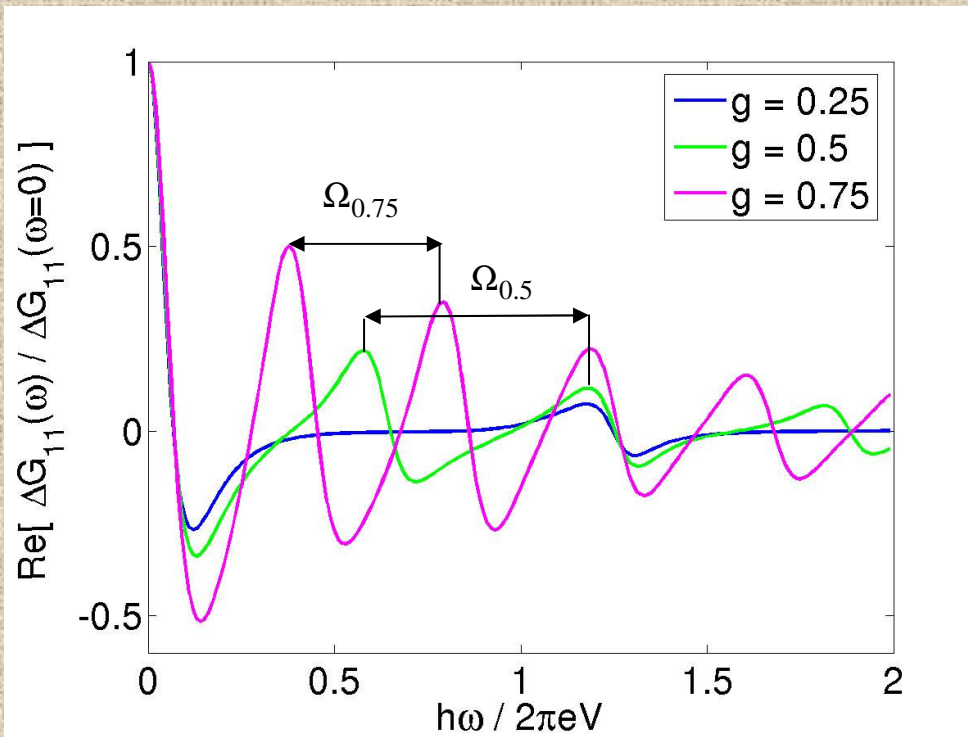
# AC CONDUCTANCE

## WEAK-BACKSCATTERING LIMIT OF THE EXCESS AC CONDUCTANCE

$$\Delta G_{11}(\omega) = G_{11}(\omega) - G_{11}(\omega)|_{V=0} \quad \text{where} \quad V = V_2 - V_1 \quad \text{source-drain voltage}$$

FOR  $g=1$ :  $\Delta G_{11}(\omega) = 0$  *because of the linearity of the I-V characteristic*

FOR  $g \neq 1$ : *oscillations with frequency with a pseudo-period related to the wire frequency*  $\omega_L = v_F / gL$



$$\begin{aligned} x_i &= 0 \\ T &= 0 \\ \lambda / eV &= 0.01 \\ g\hbar\omega_L / eV &= 0.05 \end{aligned}$$

The pseudo-period depends on  $L$  and  $g$ :

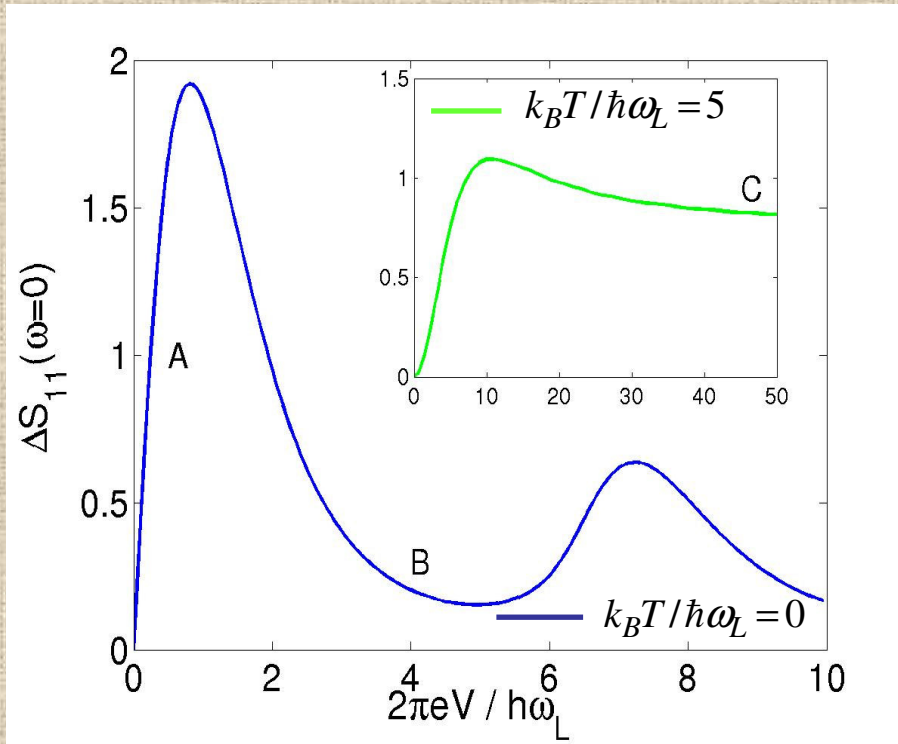
$$\Omega_g = \frac{hv_F}{gL eV}$$

# ZERO-FREQUENCY NOISE

## IN THE WEAK-BACKSCATTERING LIMIT

$$S_{nm}(\omega=0) = eI_B \coth\left(\frac{eV}{2k_B T}\right) + 2k_B T \left[ \frac{e^2}{h} - 2 \frac{\partial I_B}{\partial V} \right]$$

where  $I_B$  is the backscattering current



- **REGION A:** short-wire limit  $eV < \hbar\omega_L$   
 $\Rightarrow$  **linear variation** with voltage  
 $\Rightarrow$  qualitative agreement with experiments on carbon nanotubes

*WU et al., PRL 99, 156803 (2007)*  
*HERRMANN et al., PRL 99, 156804 (2007)*

- **REGION B:** long-wire limit  $eV > \hbar\omega_L$   
 $\Rightarrow$  **oscillations** whose envelope has a power-law dependence

- **REGION C:** high temperature limit  
 $k_B T > \hbar\omega_L$

$\Rightarrow$  behaves like the noise of an infinite length interacting wire: **power-law variation**

# FINITE-FREQUENCY NON-SYMMETRIZED NOISE

WE CALCULATE

$$\Delta S_{nm}(\omega) = S_{nm}(\omega) - S_{nm}(\omega)|_{V=0}$$

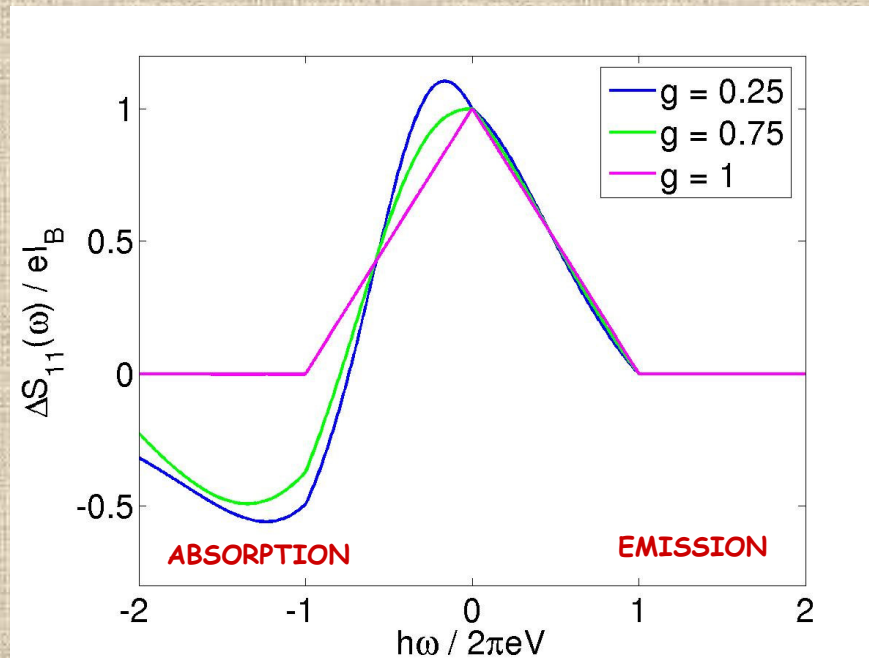
FOR  $g=1$ : the non-symmetrized excess noise is **symmetric**

$$S_{nm}(\omega) = S_{nm}^+(\omega) - \hbar\omega \text{Re}[G_{nm}(\omega)] \Rightarrow \Delta S_{nm}(\omega) = \Delta S_{nm}^+(\omega) \quad \text{because} \quad \Delta G_{11}(\omega) = 0 \quad \text{when} \quad g = 1$$

FOR  $g \neq 1$ : the non-symmetrized excess noise becomes **asymmetric**

SHORT-WIRE LIMIT

$$g\hbar\omega_L / eV = 1$$



$$x_i = 0$$

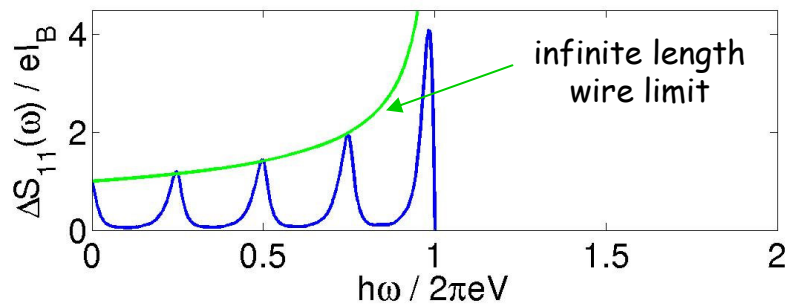
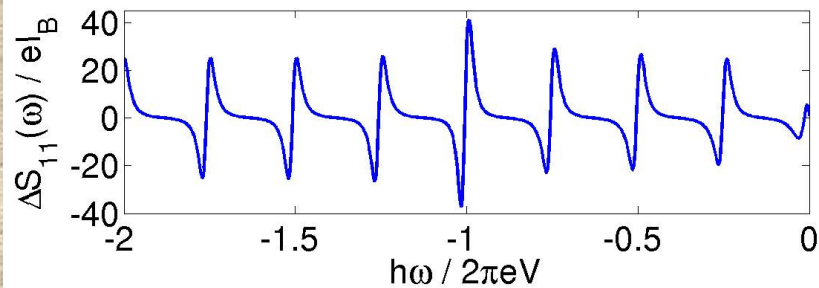
$$T = 0$$

$$\lambda / eV = 0.01$$

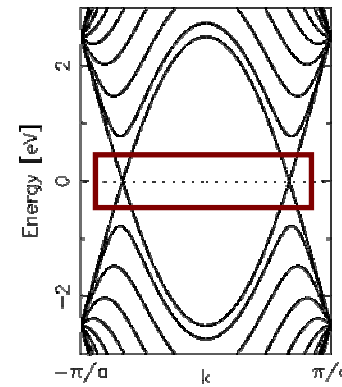
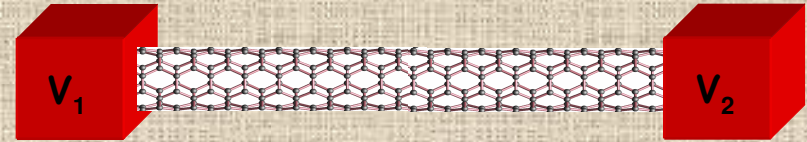
# FINITE-FREQUENCY NON-SYMMETRIZED NOISE

## LONG-WIRE LIMIT

$g = 0.25$   
 $x_i = 0$   
 $T = 0$   
 $\lambda / eV = 0.01$   
 $g\hbar\omega_L / eV = 0.01$

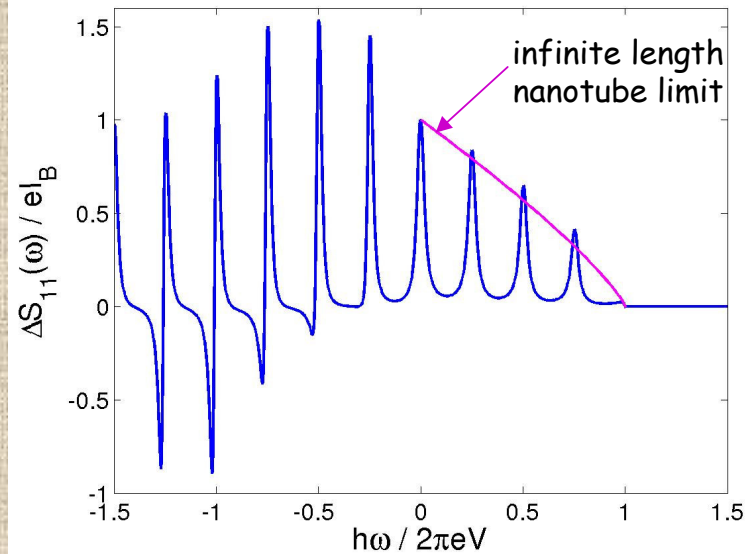


## CARBON NANOTUBE



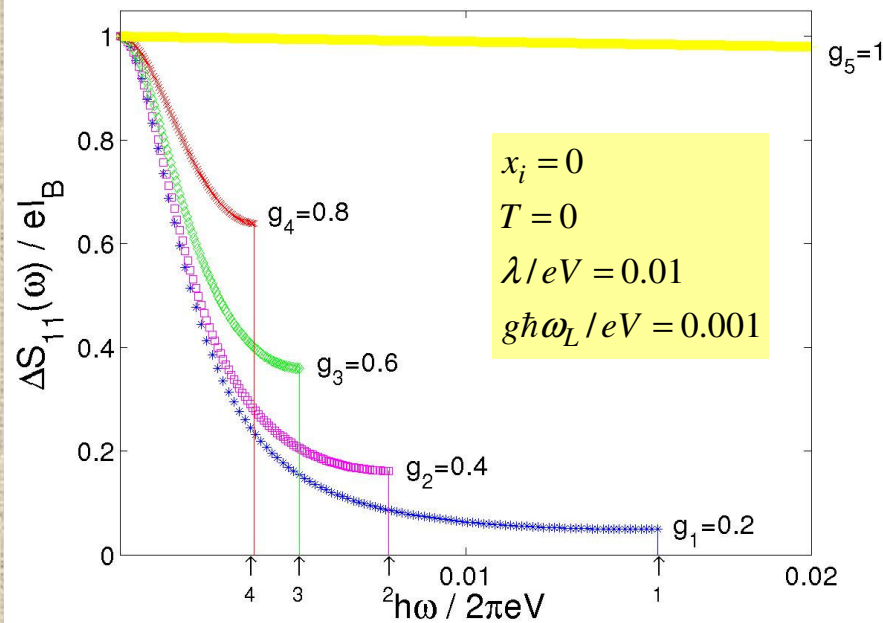
**four channels of conduction**

{ charge sector with  $g < 1$   
 { 3 others sectors with  $g = 1$

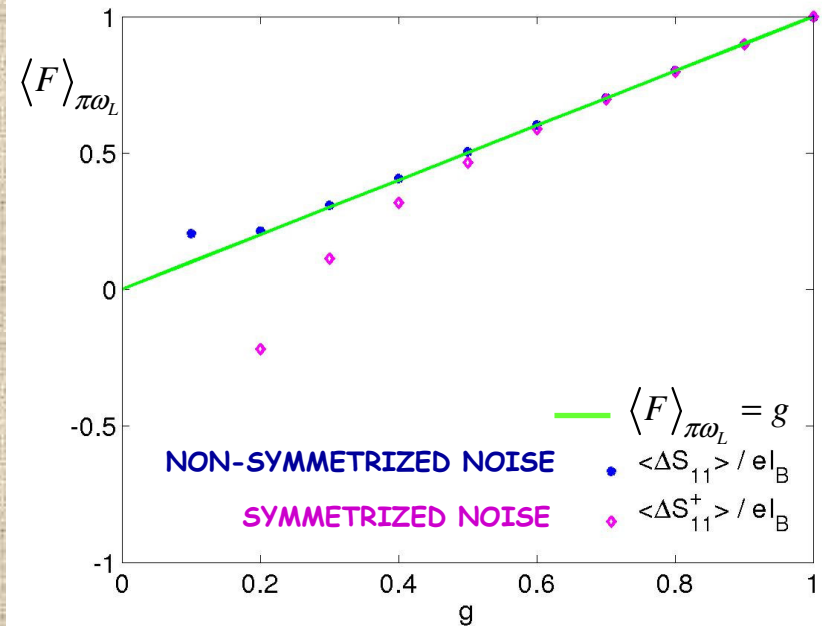


# AVERAGE NON-SYMMETRIZED NOISE

## EMISSION EXCESS NOISE



## AVERAGED FANO FACTOR



WE CALCULATE THE AVERAGE OVER THE FIRST HALF PERIOD

$$\langle F \rangle_{\pi\omega_L} = \frac{\langle \Delta S_{11}(\omega) \rangle_{\pi\omega_L}}{eI_B} \quad \text{where} \quad \langle \Delta S_{11}(\omega) \rangle_{\pi\omega_L} = \frac{1}{\pi\omega_L} \int_0^{\pi\omega_L} \Delta S_{11}(\omega) d\omega$$

⇒ we obtain  $\langle F \rangle_{\pi\omega_L} \approx g$

⇒ it should be possible to extract the value of the interaction parameter  $g$

# CONCLUSION

- **Simple relation** between the AC conductance the non-symmetrized noise

$$S_{nm}(\omega) = S_{nm}^+(\omega) - \hbar\omega \text{Re}[G_{nm}(\omega)]$$

- In the presence of Coulomb interactions, the non-symmetrized noise is asymmetric:

**Emission noise ( $\omega > 0$ )  $\neq$  Absorption noise ( $\omega < 0$ )**

- At low-temperature and for a long wire or a long nanotube, we obtain **oscillations** with a period related to  $L$  and  $g$
- The **average non-symmetrized excess noise** over the first half period gives the value of  $g$

$$\frac{\langle \Delta S_{11}(\omega) \rangle_{\pi\omega_L}}{eI_B} \approx g$$

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